## 1.2.2 CONVOLUTION

Characterization of behavior of any LTI system.

1. Decompose input into sum of simple basis functions  $x_i(n)$ .

$$\mathbf{x}(\mathbf{n}) = \sum\limits_{i=0}^{N-1} \, \mathbf{c}_i \mathbf{x}_i(\mathbf{n})$$

$$y(n) = S[x(n)]$$

$$= \sum_{i=0}^{N-1} \, c_i \, \operatorname{S}[x_i(n)]$$

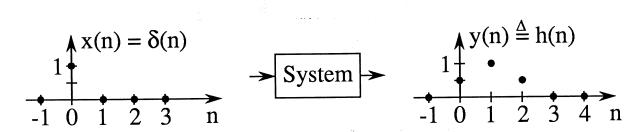
2. Choose basis functions which are all shifted versions of a single basis function  $x_0(n)$ .

$$\label{eq:i.e.} \begin{array}{ll} \textit{i.e.} & x_i(n) = x_0(n-n_i) \\ \\ \text{Let} & y_i(n) = S \; [x_i(n)], \qquad i=0,...,N-1 \\ \\ \text{then} & y_i(n) = y_0(n-n_i) \end{array}$$

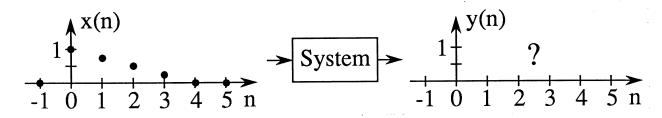
and 
$$y(n) = \sum_{i=0}^{N-1} c_i y_0(n-n_i)$$

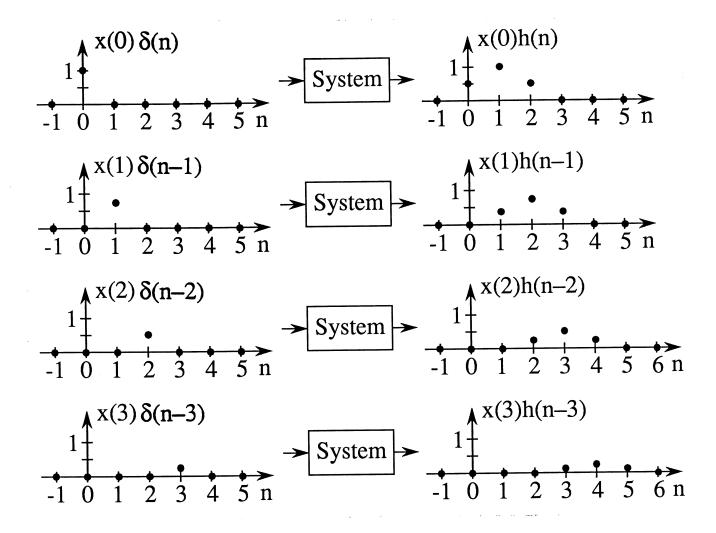
### 3. Choose impulse as basis function.

Denote impulse response by h(n)

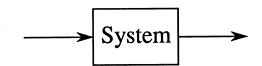


Now consider an arbitrary input x(n).

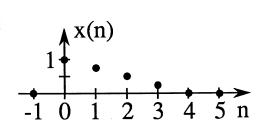


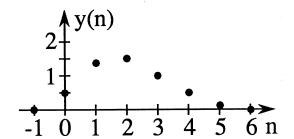


Sum over both the set of inputs and the set of outputs.



$$\begin{array}{lll} x(n) &= x(0) \, \delta(n) & y(n) &= x(0) \, h(n) \\ &+ x(1) \, \delta(n{-}1) & + x(1) \, h(n{-}1) \\ &+ x(2) \, \delta(n{-}2) & + x(2) \, h(n{-}2) \\ &+ x(3) \, \delta(n{-}3) & + x(3) \, h(n{-}3) \end{array}$$





### Convolution Sum

$$\mathbf{x}(\mathbf{n}) = \sum_{k=-\infty}^{\infty} \mathbf{x}(k) \, \delta(\mathbf{n} - \mathbf{k})$$

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$

let 
$$\ell = n - k \Rightarrow k = n - \ell$$

$$y(n) = \sum_{\ell=\infty}^{-\infty} x(n-\ell) \, h(\ell) = \sum_{\ell=-\infty}^{\infty} x(n-\ell) \, h(\ell)$$

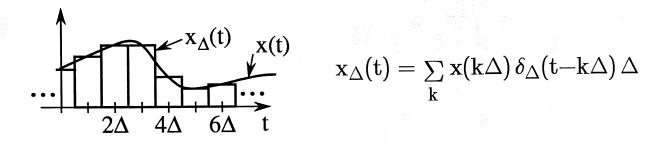
#### Notation and Identity

For any signals  $x_1(n)$  and  $x_2(n)$ , we use an asterisk to denote their convolution; and we have the following identity

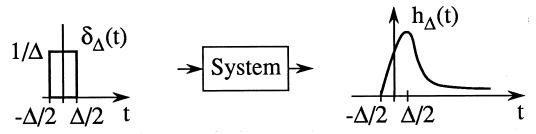
$$\begin{split} x_1(n) * x_2(n) &= \sum_{k=-\infty}^{\infty} x_1(n-k) x_2(k) \\ &= \sum_{k=-\infty}^{\infty} x_1(k) x_2(n-k) \;. \end{split}$$

## Characterization of CT LTI Systems

1. Approximate x(t) by a superposition of rectangular pulses:

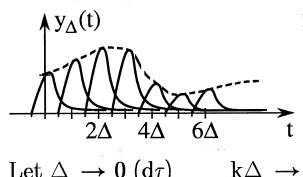


2. Find response to a single pulse:



# 3. Determine response to $x_{\Delta}(t)$ :

$$\mathbf{y}_{\Delta}(\mathbf{t}) = \sum\limits_{\mathbf{k}} \mathbf{x}(\mathbf{k}\Delta) \, \mathbf{h}_{\Delta}(\mathbf{t} - \mathbf{k}\Delta) \, \Delta$$



$$\begin{array}{lll} \operatorname{Let} \ \Delta \ \to \ 0 \ (\mathrm{d}\tau) & \mathrm{k}\Delta \ \to \ \tau \\ \delta_{\Delta}(t) \ \to \ \delta \ (t) & \mathrm{x}_{\Delta}(t) \ \to \ \mathrm{x}(t) \\ \mathrm{h}_{\Delta}(t) \ \to \ \mathrm{h} \ (t) & \mathrm{y}_{\Delta}(t) \ \to \ \mathrm{y}(t) \end{array}$$

# Convolution Integral

$$\mathbf{x}(\mathbf{t}) = \int_{-\infty}^{\infty} \mathbf{x}(\tau) \, \delta(\mathbf{t} - \tau) d\tau$$

$$y(t) = \int\limits_{-\infty}^{\infty} x(\tau) h(t-\tau) d au$$

#### Notation and Identity

For any signals  $x_1(t)$  and  $x_2(t)$ , we use an asterisk to denote their convolution; and we have the following identity

$$\begin{split} \mathbf{x}_1(t) * \mathbf{x}_2(t) &= \int\limits_{-\infty}^{\infty} \mathbf{x}_1(\tau) \mathbf{x}_2(t-\tau) \mathrm{d}\tau \\ &= \int\limits_{-\infty}^{\infty} \mathbf{x}_1(t-\tau) \mathbf{x}_2(\tau) \mathrm{d}\tau \;. \end{split}$$

#### Example:

DT System 
$$y(n) = \frac{1}{W} \sum_{k=0}^{W-1} x(n-k)$$
 W - integer

Find response to  $x(n) = e^{-n/D}u(n)$ 

W - width of averaging window

D – duration of input

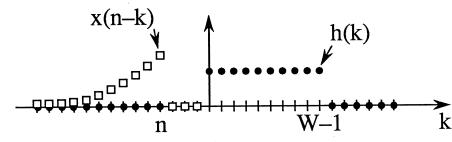
To find impulse response, let  $x(n) = \delta(n) \Rightarrow h(n) = y(n)$ 

$$h(n) = \frac{1}{W} \sum_{k=0}^{W-1} \delta(n-k) = \begin{cases} 1/W, & 0 \le n \le W-1 \\ 0, & else \end{cases}$$

Now use convolution to find response to  $x(n) = e^{-n/D}u(n)$ .

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k) h(k)$$

Case 1: n < 0



$$y(n) = 0$$

#### Case 2: $0 \le n \le W-1$

$$y(n) = \sum_{k=0}^{n} x(n-k) h(k)$$

$$y(n) = \sum_{k=0}^{n} x(n-k) h(k)$$

$$= \frac{1}{W} \sum_{k=0}^{n} e^{-(n-k)/D}$$

$$= \frac{1}{W} e^{-n/D} \sum_{k=0}^{n} e^{k/D}$$

# Geometric Series

$$\sum\limits_{k=0}^{N-1}\,z^k=\frac{1-z^N}{1-z}$$
 ,  $\;\;$  for any complex number  $z$ 

$$\label{eq:continuous_signal} \begin{array}{l} \sum\limits_{k=0}^{\infty} \, z^k = \frac{1}{1-z} \; , \quad \mid z \mid \; < 1 \end{array}$$

$$y(n) = \frac{1}{W} e^{-n/D} \left[ \frac{1 - e^{(n+1)/D}}{1 - e^{1/D}} \right]$$

$$= \frac{1}{W} \left[ \frac{1 - e^{-(n+1)/D}}{1 - e^{-1/D}} \right]$$

Case 3:  $W \leq n$ .

$$h(k) = x(n-k)$$

$$h(k) = x(n-k)$$

$$W-1 = x(n-k)$$

$$y(n) = \sum_{k=0}^{W-1} x(n-k)h(k)$$

$$= \frac{1}{W} \sum_{k=0}^{W-1} e^{-(n-k)/D}$$

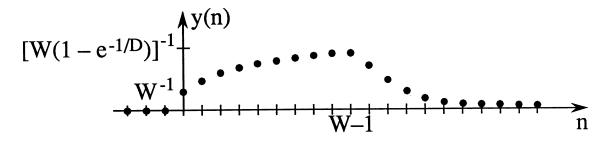
$$= \frac{1}{W} e^{-n/D} \sum_{k=0}^{W-1} e^{k/D}$$

$$y(n) = \frac{1}{W} e^{-n/D} \left[ \frac{1 - e^{W/D}}{1 - e^{1/D}} \right]$$

$$= \frac{1}{W} \left[ \frac{1 - e^{-W/D}}{1 - e^{-1/D}} \right] e^{-[n-(W-1)]/D}$$

## Putting everything together

$$y(n) = \begin{cases} 0, & n < 0 \\ \frac{1}{W} \left[ \frac{1 - e^{-(n+1)/D}}{1 - e^{-1/D}} \right], & 0 \le n \le W - 1 \\ \frac{1}{W} \left[ \frac{1 - e^{-W/D}}{1 - e^{-1/D}} \right] e^{-[n - (W-1)]/D}, & W \le n \end{cases}$$



#### Causality for LTI Systems

$$\begin{array}{c} y(n) = \sum\limits_{k=-\infty}^{n} x(k)\,h(n-k) + \sum\limits_{k=n+1}^{\infty} x(k)\,h(n-k) \\ \text{contribution} & \text{contribution} \\ \text{from past and} & \text{from future} \\ \text{present inputs} & \text{inputs} \end{array}$$

System will be causal  $\Leftrightarrow$  second sum is zero for any input x(k).

This will be true 
$$\Leftrightarrow h(n-k) = 0, k = n+1,...,\infty$$
 
$$\Leftrightarrow h(k) = 0, k < 0$$

... A LTI system is  $causal \Leftrightarrow h(k) = 0$ , k < 0, i.e. the impulse response is a  $causal \ signal$ 

#### Stability for LTI Systems

Suppose the input is bounded, i.e.  $M_x < \infty$ .

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$

$$|y(n)| = |\sum_{k=-\infty}^{\infty} x(k) h(n-k)|$$

$$\leq \sum_{k=-\infty} |x(k)| |h(n-k)|$$

$$= M_x \sum_{k=-\infty}^{\infty} |h(k)|$$

$$= -\infty$$

... It is sufficient for BIBO stability that the impulse response be absolutely summable.

Is it necessary?

Suppose 
$$\sum_{k} |h(k)| < \infty$$

Consider 
$$y(0) = \sum_{k} x(k) h(-k)$$

Assuming h(k) is real-valued, let

$$\mathbf{x}(\mathbf{k}) = \begin{cases} 1, & \mathbf{h}(-\mathbf{k}) > 0 \\ -1, & \mathbf{h}(-\mathbf{k}) < 0 \end{cases}$$

then 
$$y(0) = \sum_{k} |h(k)| < \infty$$

... A LTI system is BIBO stable  $\Leftrightarrow$  the impulse response is absolutely summable, i.e.  $\sum\limits_{\mathbf{k}} |\mathbf{h}(\mathbf{k})| < \infty$ 

### Example

$$y(n) = x(n) + y(n-1)$$

Find the impulse response.

Let 
$$x(n) = \delta(n)$$
, then  $h(n) = y(n)$ 

Need to find solution to

$$y(n) = \delta(n) + 2y(n-1)$$

This example differs from earlier ones because the system is recursive, *i.e.* the current output depends on previous output values as well as the current and previous inputs.

- 1. must specify initial conditions for the system (assume y(-1) = 0).
- 2. cannot directly write a closed form expression for y(n).

Find output sequence term by term

$$y(0) = \delta(0) + 2y(-1) = 1 + 2(0) = 1$$
 $y(1) = \delta(1) + 2y(0) = 0 + 2(1) = 2$ 
 $y(2) = \delta(2) + 2y(1) = 0 + 2(2) = 4$ 

Recognize general form

$$h(n)=y(n)=2^nu(n)$$

- 1. Assuming system is initially at rest, it is causal.
- 2.  $\sum_{n} |h(n)| < \infty \Rightarrow \text{system is not BIBO stable.}$